

[0035] The various codes 106 are then used to modulate or otherwise create the particular modulation signals 110A-D that are applied to the various sensing electrodes 112A-D in sensing region 101 (step 202). As described above, the applied signals are electrically affected by the presence of object 121, with the resultant electrical effects being determinable from received signal 116 (step 203).

[0036] Demodulating received signal channels 113 (in step 204) suitably involves extracting information about the position of object 121 from the modulated signals. Such extraction typically involves reversing the modulation process described above. Accordingly, demodulator 117 typically receives carrier signal 111 (or another signal that is synchronous with signal 111) for performing analog demodulation and/or signal discrimination (e.g. distinguishing between noise and desired signal) in addition to the particular digital code 106 that previously modulated the carrier signal used to create the particular resultant signal 116. Because the sensor both transmits and receives, it is rarely necessary to recover the carrier or code sequence.

[0037] Demodulation may be carried out for any number of received signal channels 113, as appropriate (step 206). In the exemplary sensor 100 shown in FIG. 1A, signal 116 resulting from the transmission of each signal channel from the modulated electrodes 112A-D are received on a common path emanating from receiving electrode 114. Even if the various sensing channels 113A-D are all active at the same time (e.g. modulation signals 110A-D are simultaneously provided to each modulated electrode 112A-D), however, the resulting signals 116 produced by each channel 113A-D can be demodulated using conventional CDM demodulation techniques. Particular components (or channels) of resultant signal 116 produced in response to any modulation signal 110A-D can therefore be readily extracted. This concept can be exploited in myriad ways, as described below, to create a number of additional features and performance enhancements within sensor 100. A common modulation signal 110A-D, for example, could be applied to multiple electrodes 112A-D to increase the size of any particular sensing zone within region 101. These zones can be readily adjusted during operation to create various operating modes or the like. To make the entire sensing region 101 act as a single button, for example, each electrode 112A-D could be provided with the same modulation signal 110 without otherwise adjusting the performance of the sensor. Because all of the signals resulting from receive electrode 116 are provided on a common path in FIG. 1, simply demodulating the entire received signal using the common modulation code will identify the presence of object 121 anywhere within sensing region 101 in this case. Similar concepts can be applied to create any number of independent or overlapping sensing zones across sensing region 101 through simple manipulation of digital code sequences 106. Furthermore, spatial frequency filtering can be done simply through proper modulation and demodulation, for example to accomplish palm rejection or to reject other inappropriate inputs.

[0038] The demodulated signals 118 are appropriately received at controller 102 so that the position-based attribute of object 121 can be determined (step 208). These signals may be filtered digitally or as analog, using linear and non-linear filters. Various techniques for identifying the position of object 121 with respect to the various electrodes 112A-D include detection of peak electrical effect, compu-

tation of a centroid based upon the electrical effects, comparison of differences in electrical effects observed between electrodes 112A-D, comparison of changes in electrical effects over time, interpolation between signal channels from the electrodes, and/or according to many other techniques. In the case of peak detection, the position of object 121 is associated with one or more electrodes 112A-D by identifying which modulation signal 110A-D produced the greatest relative change of capacitive effect in resultant signal 116. Sensing channels 113A-D experiencing such peak (maximum, minimum, or otherwise distinct) electrical effects could also be identified by comparing currently-observed, scaled electrical effects to baseline values (e.g. average values for the particular channel that are empirically determined, averaged over time, stored from a previous observation, and/or the like). Still other embodiments could identify the channel 113A-D that produced peak electrical effects by comparing current electrical effects for each channel 113A-D with current values observed in neighboring sensing channels. Alternatively, a weighted average of the electrical effects observed from some or all of the modulated electrodes 112A-D can be computed, with this weighted average, or centroid, correlating to the position of object 121. Many techniques for correlating electrical effects observed across sensing region 101 to a position of object 121 are known or may be subsequently developed, and any of these techniques may be used in various embodiments, according to the application.

[0039] By varying the digital codes 106 used to create modulation signals 110A-D over time, various additional features can be implemented. To implement a simple dual-differential digital-to-analog conversion for received signal channels 113, for example, the digital code 106 applied to one or more electrodes 112 is logically inverted (e.g. 1's complement) on a periodic, aperiodic, or other time basis to obtain complementary sensed signals 116. The complementary codes 106 can be used to drive two separate ADC inputs (e.g. ADCs present in driver 115 and/or demodulator 117) in opposite directions, thereby canceling out many types of variability or residual distortion in signal 116. Steps 210 and 212 describe optional noise reconfiguration and image processing features, respectively, that may be enabled in various embodiments as additional benefits available from the use of spread spectrum techniques. These features are described in increasing detail below (in conjunction with FIGS. 3 and 4, respectively), and may not be present within all embodiments. Because the digital codes 106 are inherently simple to modify, store and subsequently process, any number of signal enhancement, noise reduction and/or other performance improvements to sensor 100 are enabled. Further, a relatively large number of digital codes are available due to the combinatorial power of digital sequences. Coding gain and orthogonality conventionally rely upon linearity and superposition of the particular codes used. Although non-linearity and dispersion limit the theoretical effectiveness of digital codes, these limitations can be more than offset by the increases in relative signal power (and thus SNR) that can result from simultaneously modulating more than one electrode 112. Further, since it is possible that self-induced inter-channel noise dominates over other noise sources in many embodiments, a relatively stable dynamic range can be provided.

[0040] Referring now to FIGS. 3A-B, spread spectrum techniques allow for improved noise avoidance as compared